Antennas for cellular base stations

— challenges, trends and constraints —

by Jan Hesselbarth, University of Stuttgart

FP7–ARTISAN meeting, Belfast, January 30, 2014

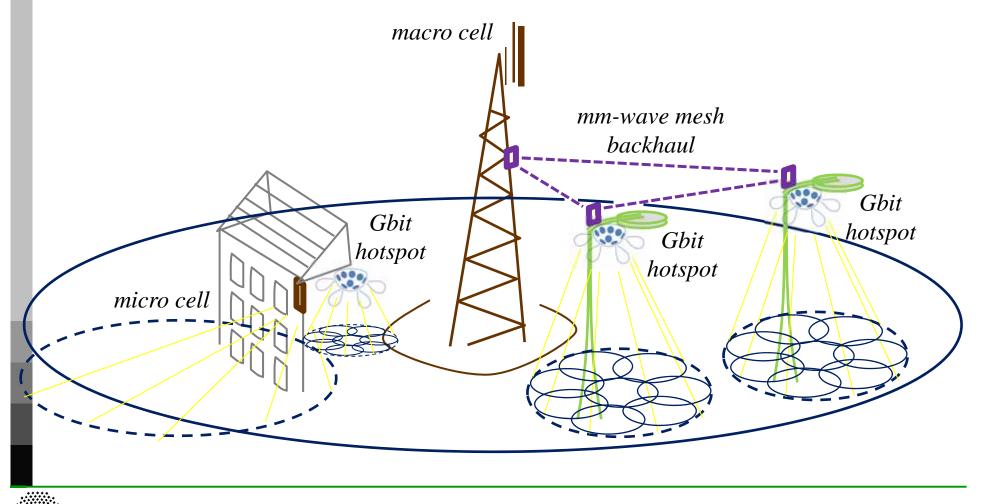
outline:

- choice of frequency
- antenna radiator types and characteristics
- macro sector antennas
- antennas for in-buiding and in-cabin systems
- wireless backhaul
- what's next



# Antennas to provide coverage, throughput, adaptivity

Different requirements for frequency, pattern, adaptivity, size, cost etc.



#### outline:

### - choice of frequency

- antenna radiator types and characteristics
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# **Choice of frequency**

Criteria for mobile cellular:

antenna size, path loss, diffraction, bandwidth, Doppler

#### low frequency cellular:

- low path loss / large cells
- strong diffraction
- small Doppler shift
- small bandwidth
- large antennas
- $\rightarrow$  for coverage and mobility
- $\rightarrow$  small throughput

```
(450 MHz) & 700 – 1000 MHz
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### high frequency cellular:

- high path loss / small cells
- weak diffraction
- large Doppler shift
- larger bandwidth
- smaller antennas
- → for increased throughput in rather small cells

#### 1700 – 2600 MHz

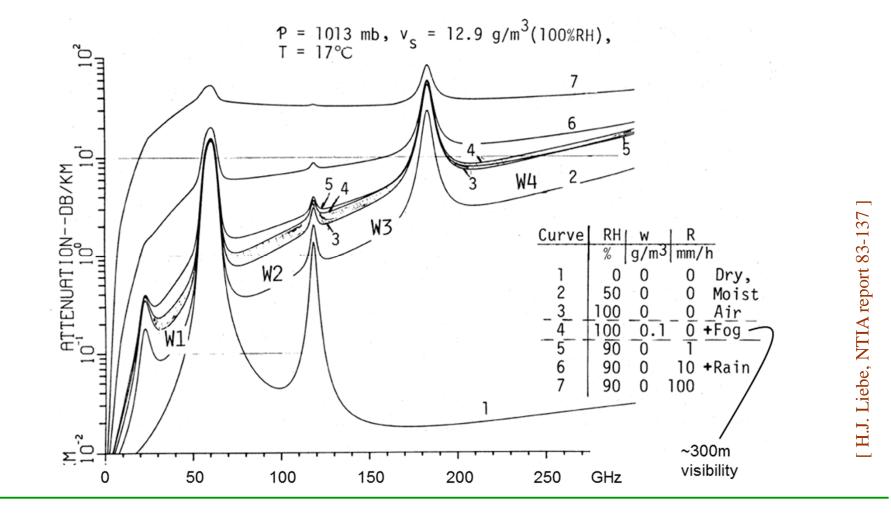
Example: "NMT" network in Skandinavia at 450 MHz: 25 km cell radius, few users → congested in some cities as early as 1983



## **Choice of frequency**

Criteria for wireless backhaul:

path loss, bandwidth, atmospheric attenuation, licensing scheme

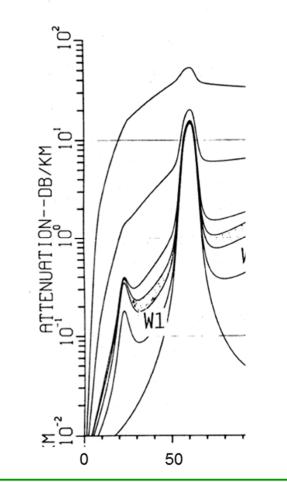




# **Choice of frequency**

#### Criteria for wireless backhaul:

path loss, bandwidth, atmospheric attenuation, licensing scheme



- → long-distance hops (>10 km) with raher low capacity at 6...20 GHz . dish diameter ~ 1...2 m
- → medium-distance hops (3...10 km) with high capacity (~ 100 Mbps) at 28...44 GHz . dish diameter ~ 0.6 m
- → shortest-distance hops (<1km) with multi-GBps capacity at 60 GHz
  . dish diameter ~ 0.3m
- → short-distance hops (<2km) with multi-GBps capacity at 71...86 GHz . dish diameter ~ 0.3m



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#### outline:

- choice of frequency

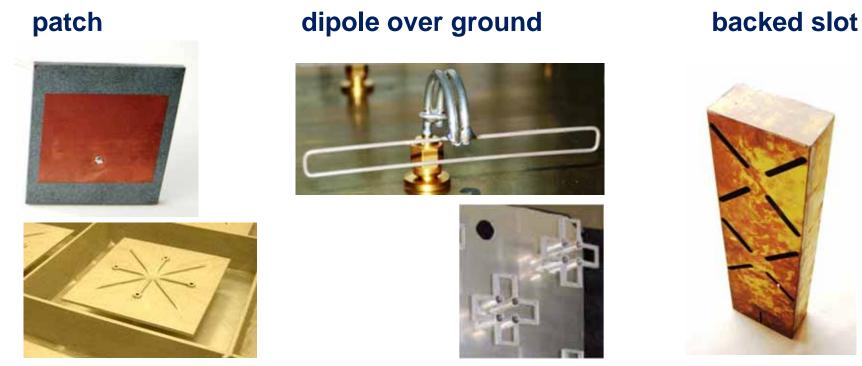
### - antenna radiator types and characteristics

- macro sector antennas
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For most antennas with sectorial pattern (not: omnidirectional ones), a groundplane provides suppression of backward radiation and is used for mounting purposes.

Radiators with groundplane:



#### ... all similar in gain (5...9 dBi) and in beamwidth (90°...140°)

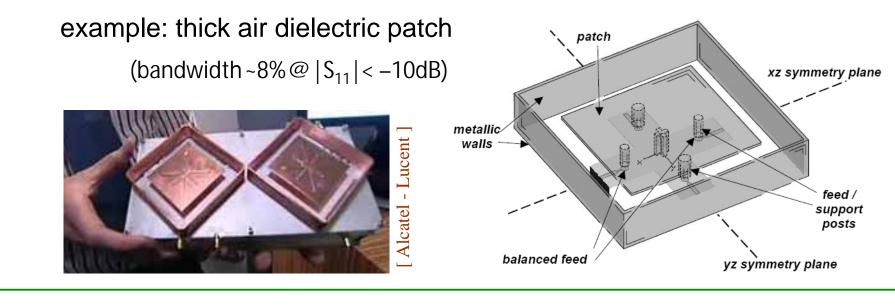


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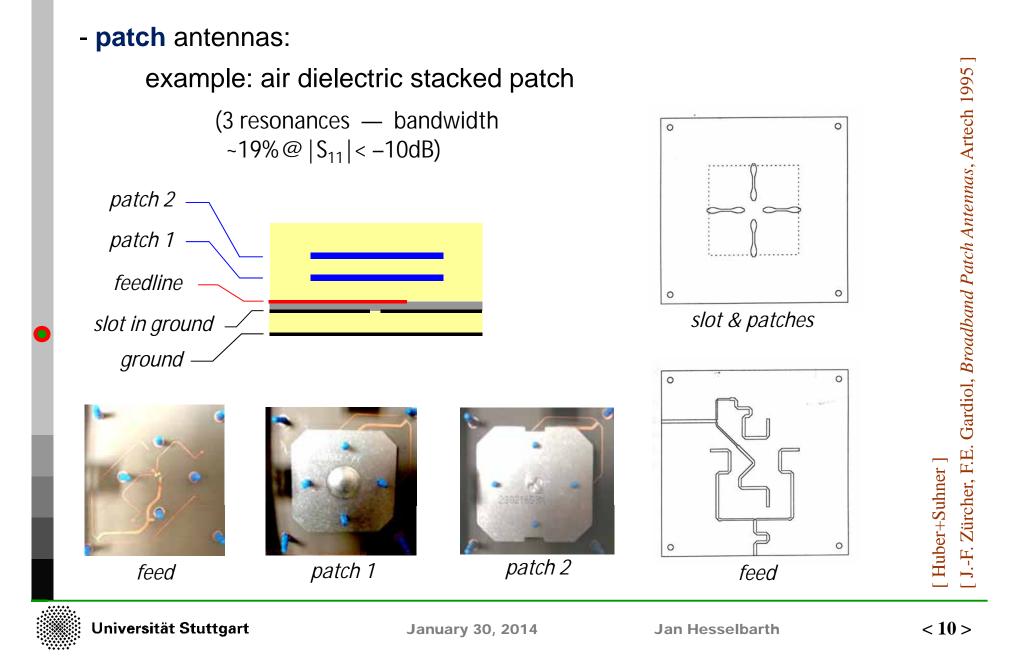
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- backed slot antennas: severe problems with bandwidth
→ no relevant use

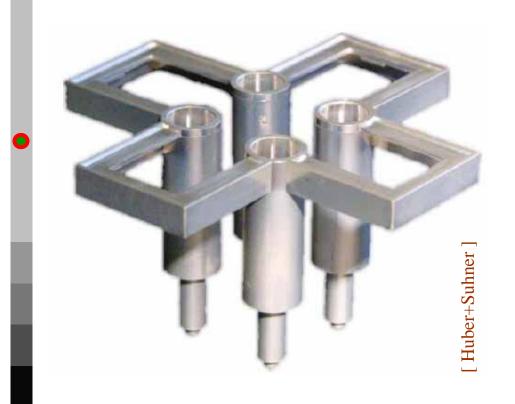
 patch antennas: reasonable bandwidth requires very thick dielectric or stacked patches; efficiency and cost and weight requirements lead to mechanically tricky air dielectric
→ limited use

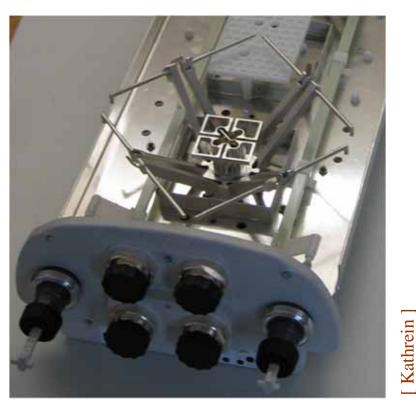






- dipole-over-ground: good radiator bandwdith, wideband balun needed, rather thick, various low-cost 3D technologies possible (punched sheet metal, circuit board arrangement, metalized molded plastic), many PIM-critical connections
→ widest use







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### Macro sector antennas

- a column of (almost-) in-phase radiators

wide horizonthal (azimuth) beam pattern (3dB BW ~  $60^{\circ}$  ... 120°) focused vertical (elevation) beam pattern (3dB BW ~  $6^{\circ}...12^{\circ}$ ) linear array of  $\lambda/2$ ,  $\lambda/4$ -over-ground dipoles; equal-magnitude, equalphase,  $0.8\lambda$  spacing: maximum acceptable spacing is 0.8...0.9  $\lambda$ 6 elements: because of sidelobe 15.3 dBi, SLL -13.2 dB, level 3dB BW 10.5° 10 elements: 17.5 dBi, SLL -13.3 dB, 3dB BW 6.3°



### Macro sector antennas

- single (V) polarization rarely used
- standard is dual (+45°/–45°) polarization for diversity receive
- "broadband" antennas cover about 20% bandwidth



[Huber+Suhner]

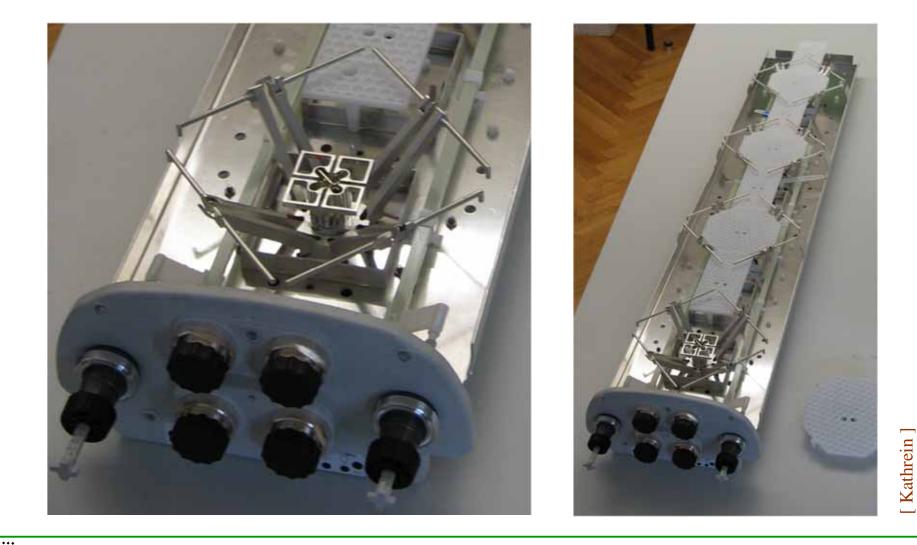




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### Macro sector antennas

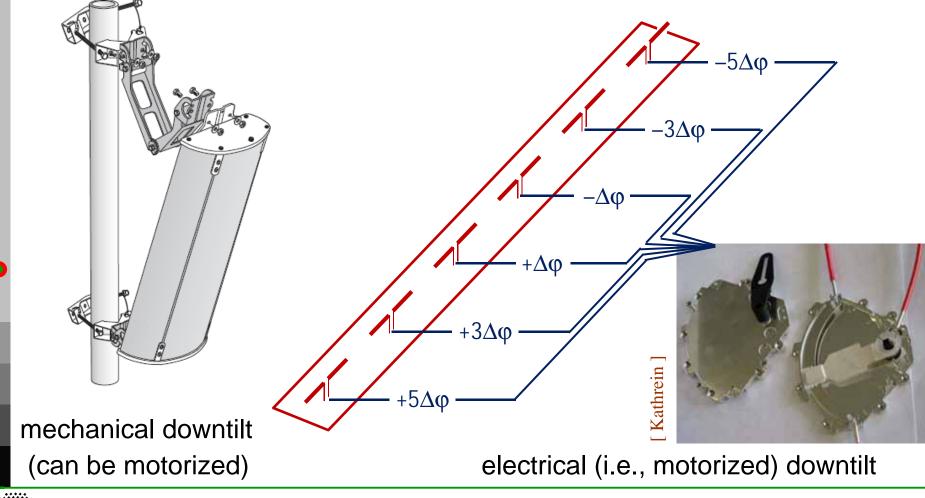
- "multi-band" antennas have separate radiator columns with separate feeds





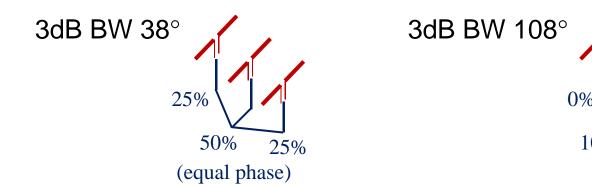
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- adaptive downtilt adapts the max cell radius and/or cell edge coverage



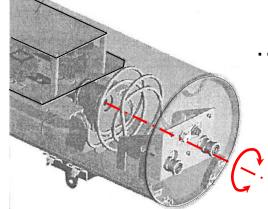


- adaptive beamwidth adapts cell sector width (e.g., 3dB BW 35°...105°)



 $\dots$  using a motorized differential phase shifter and a 90° hybrid

- adaptive pan (azimuth beam steering by, e.g.,  $\pm 30^{\circ}$ )



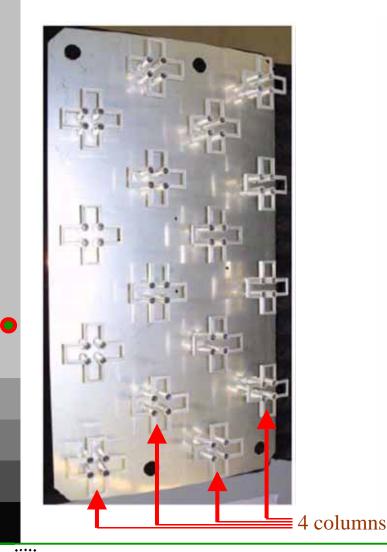
.. motorized rotation of the complete radiator column inside the radome antenna box



100%

0%

#### - beam steering / switching using multi-column antennas



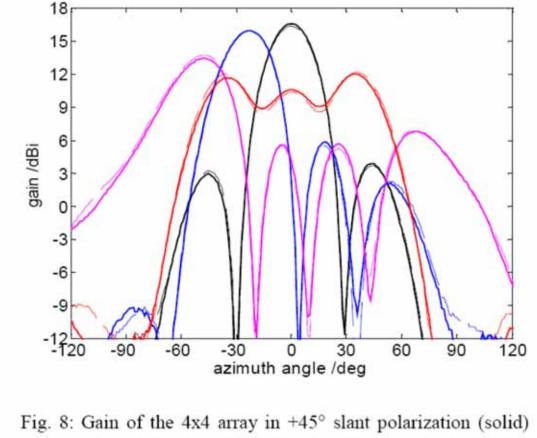
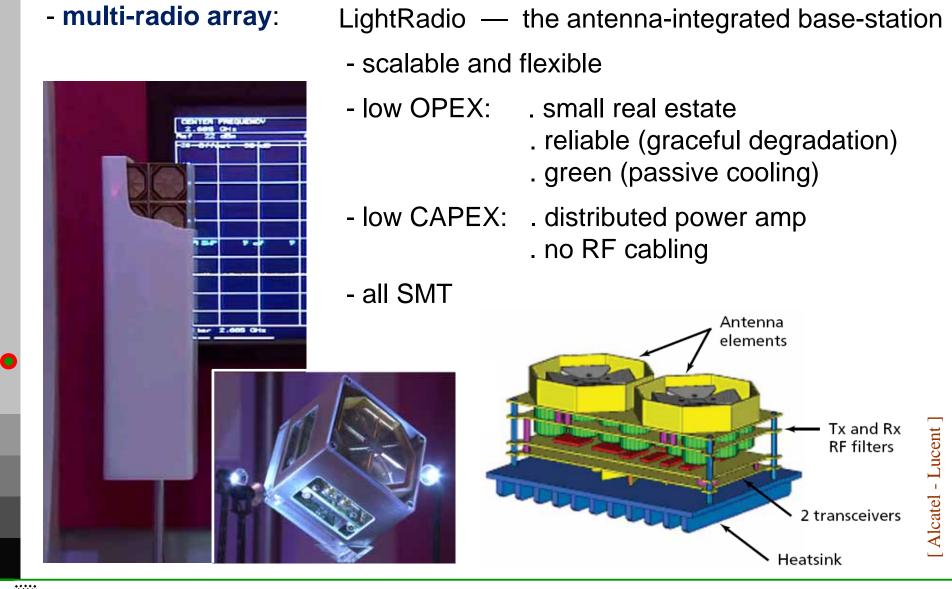


Fig. 8: Gain of the 4x4 array in  $+45^{\circ}$  slant polarization (solid) and  $-45^{\circ}$  polarization (dashed) for inphase feed (black),  $0^{\circ}/75^{\circ}/150^{\circ}/225^{\circ}$  feed (blue),  $0^{\circ}/150^{\circ}/300^{\circ}/450^{\circ}$  feed (magenta),  $0^{\circ}/120^{\circ}/120^{\circ}/0^{\circ}$  feed (red), respectively, at 2 GHz. Equal power.

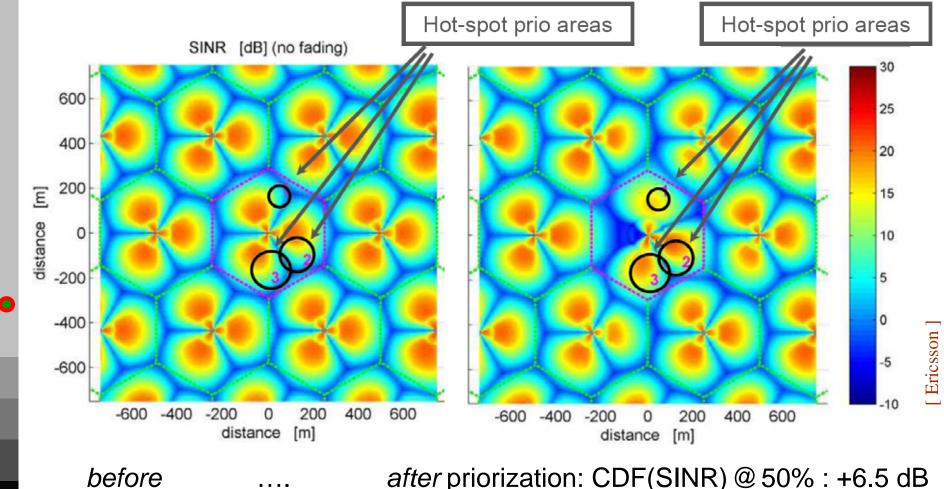


[Huber+Suhner





### - application of adaptivity: priorization of hot-spot areas



after priorization: CDF(SINR) @ 50% : +6.5 dB



# Additional aspects: weight, modularity

- weight: arrays and antennas with active electronics quickly become heavy
  - . metalized plastics instead of metal
  - . carbon structures for frames and support
  - . sandwich structures for maximum stability



[ Huber+Suhner ]

 modularity: antennas with active electronics must be modular (repair / parts replacement without taken down from tower)



# Additional aspects: passive intermodulation — PIM

- **PIM**: in channelized FDD systems, odd-order IM of two transmit signals can mask a weak receive signal

example: GSM 1900 (US): UL @ 1850–1910 & DL @ 1930–1990 TX<sub>1</sub>=1940 MHz, TX<sub>2</sub>=1980 MHz  $\rightarrow$  IM<sub>2TX1-TX2</sub>=1900 MHz

given the receive sensitity, acceptable PIM level must be very small

standard test uses two signals of +43 dBm measured PIM is at -100...-120 dBm, that is, up to -160 dBc !!!

typical specified PIM level of a base station antenna is "<-150 dBc"

PIM measurement & calibration equipment is tricky & expensive



# Additional aspects: PIM

Any electrical non-linearity can (Murphy's law: will) cause PIM:

- . micro-flashes
  - loose metal-to-metal or oxide-to-metal joints
    - $\rightarrow$  avoid cracks in solder joints or cold solder joints
    - $\rightarrow$  avoid loosened screws, bolts and connectors
    - $\rightarrow$  clean production avoid metal dust in the device
- . non-linear material
  - even some polymers produce PIM
    - $\rightarrow$  PTFE, PE do not
  - corroded metal: e.g., CuO is a known semiconductor
    - $\rightarrow$  completely (!) plated metal with Sn, Ag, Au
  - magnetic material, stainless steel, Co, Ni ...
    - $\rightarrow$  avoid galvanic finish (and even PCB) with Ni adhesion layer
    - $\rightarrow$  use special galvanic processes
    - → low-PIM PCB
  - loose metallic building roof installations
    - $\rightarrow$  place antennas at roof edges



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### - antennas for in-buiding and in-cabin systems

- wireless backhaul
- what's next

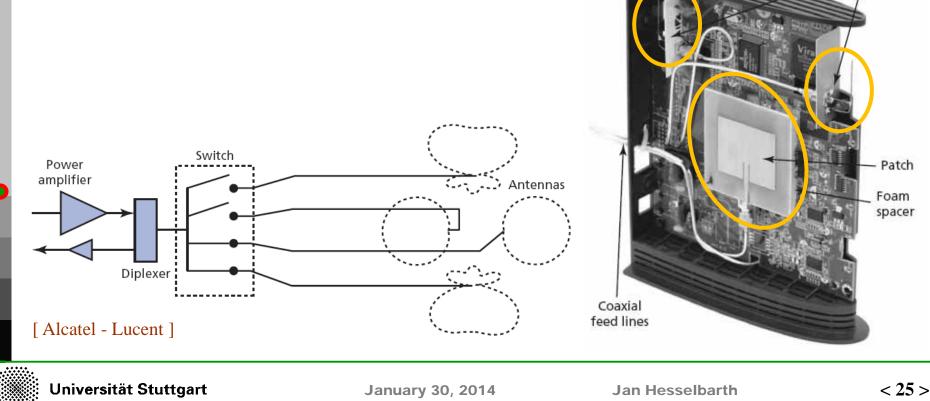


# Antennas for in-building and in-cabin systems

#### **Femto-cells**

- signalling / overlapping cells (e.g.: indoors versus outdoors) require dynamically (e.g., switched) optimized coverage using multiple antennas
- antennas must be cheap;





IFA

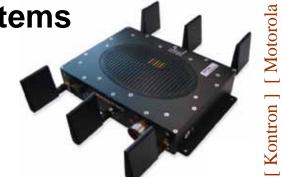
# Antennas for in-building and in-cabin systems

**in-cabin WiFi** (also: GSM1800) does not benefit from directive antennas

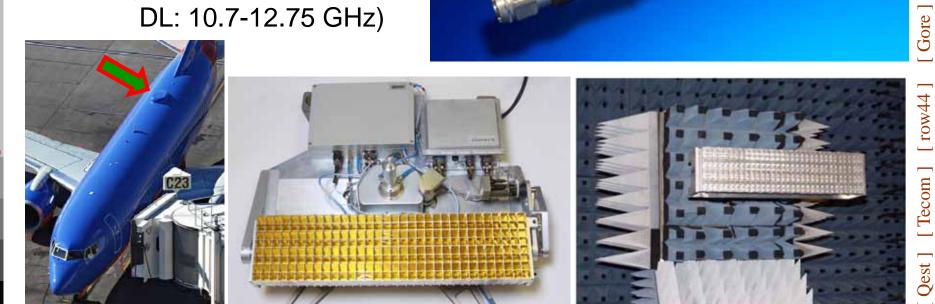
several hotspots will guarantee complete coverage

alternative: leaky cable "antenna"

**backhaul** (aircraft-to-satellite): Ku band (UL: 13.75-14.5 GHz, DL: 10.7-12.75 GHz)









[SITA OnAir

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# Wireless backhaul

LTE-tower macro cell backhaul is >500 MBps gross: optical fiber or mm-wave small cell backhaul can be anything "small", including in-band or copper wire

- wired acces (optical, copper) is preferred if existent (of non-existent, it is often too time-consuming and/or too expensive to be built)
- in-band backhaul is a waste of precious (0.7 GHz 3 GHz) frequency
- microwave (6 GHz 20 GHz) allows long distance (~ 20 km) but needs large dishes and has problems with datarates > 100 MBps
- Ka-band (28 GHz 44 GHz) is preferred for macro cell but urban areas may run out of capacity
- 60 GHz (59 GHz 64 GHz) becomes the best solution for dense deployments of small cells and fast/ non-permanent installations
- E-band (71 GHz 86 GHz) becomes the best and only solution for carrier-grade wireless backhaul with >> 1 GBps speed



# The parabolic dish

macro cell & small cell backhaul

- useful formulas:



farfield distance:farfield distance 
$$\approx \frac{2 (\text{aperture diameter})^2}{\lambda}$$
Friis: $D \approx G = \eta_{aperture\_efficiency} \frac{4\pi}{\lambda^2} \cdot \text{physical area}$ aperture efficiency: $\eta_{aperture\_efficiency} = \frac{\text{maximum effective area}}{\text{physical area}}$ Kraus: $D \approx G = \frac{41253}{\theta_{3dB\_beamwidth}} \Big|_{deg}$ Tai & Pereira: $D \approx G = \frac{36408}{\theta_{3dB\_beamwidth}} \Big|_{deg}$ 



# The parabolic dish

macro cell & small cell backhaul

- some typical E-band Cassegrain dishes:



1	diameter	datasheet	values	maximum		aperture		directivity from HPBW		
	[mm]	gain [dBi] HPBW		directivity [dBi]		efficiency		Kraus	Tai & Pereira	
				60 GHz	/ 90 GHz	60 GHz	/ 90 GH	z [dBi]	[dBi]	
	200	39.9	1.3°	42.0	45.6	62%	27%	43.9	43.3	-1]
	300	43.5	0.9°	45.5	49.1	63%	28%	47.1	46.5	Elva-1
	450	46.6	0.6°	49.0	52.6	58%	25%	50.6	50.0	

- narrow beamwidth requires very accurate alignment and structural stability

- radome loss ranges from nothing at a few GHz to 0.7 dB at E-band



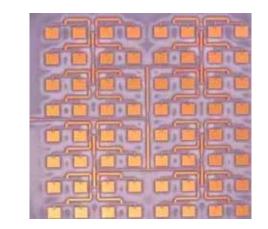
# Planar array versus parabolic dish

Can a planar array replace the parabolic dish?

- con: the parabolic mirror is a 3D structure and "looks like an antenna"
- pro: the parabolic mirror is dual polarized and has very low loss

A large planar array (32x32 or 64x64 elements)

- is flat and square and has larger aperture efficiency (> 90%) than a dish
- loss of the array is in the feed network
- most arrays are single (linearly) polarized



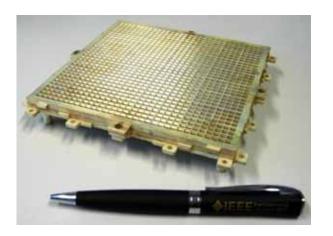
1% efficiency (~ 20 dB loss) of a 32x32 patch array with microstrip feed network at 60 GHz.

[M. Al Henawy, M. Schneider, "Planar antenna arrays at 60 GHz realized on a new thermoplastic polymer substrate," Proc. EuCAP 2010]



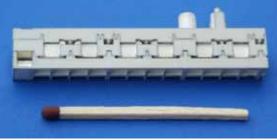
# Planar array versus parabolic dish

Can a planar array replace the parabolic dish ?



~70% efficiency (1.5 dB loss) of a 32x32 openended waveguide array with ridge waveguide feed network at 60 GHz.

[Huber+Suhner]

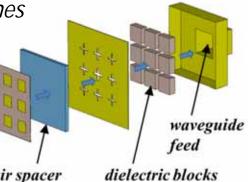


A suspended-substrate slot-coupled square patch array with waveguide feed – two feeder trees for two orthogonal linear polarization – combines



all advantages

freq-scaled array building block air spacer





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## Beam steering for mm-wave wireless backhaul

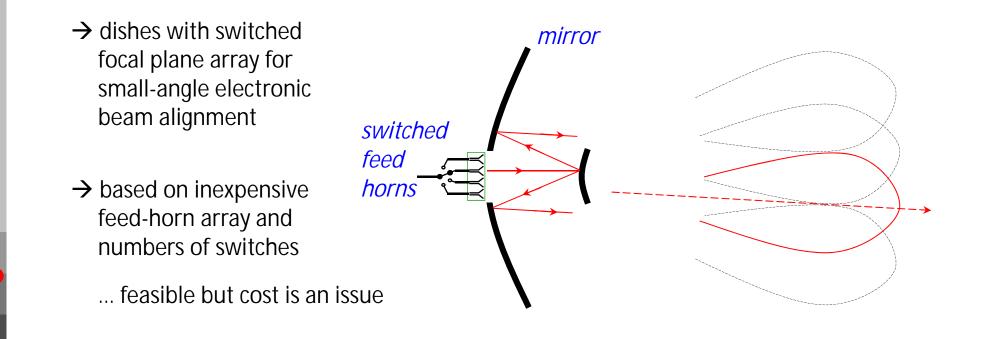
- small-angle electronic beam steering for ease of alignment alignment at installation is very expensive (mostly labor costs) automatic re-alignment would allow for reduced structural stability
- wide-angle electronic beam steering for meshed backhaul network increase of reliability and throughput

same cost reductions as small-angle alignment macro cell bit hotspot micro cell bit hotspot bit hotspot



# Beam steering for mm-wave wireless backhaul

- any possible solution must provide (reasonably) low cost and low loss
- small-angle electronic beam steering:



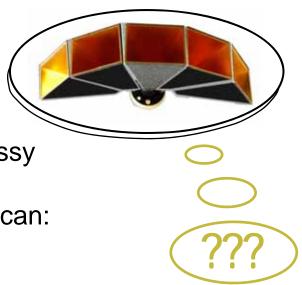
### Beam steering for mm-wave wireless backhaul

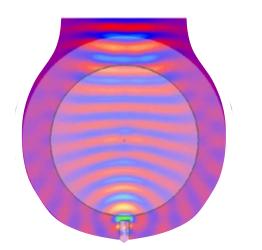
- wide-angle electronic beam steering:

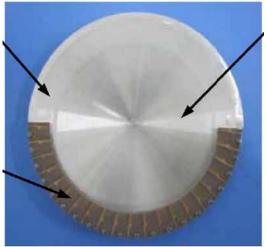
phased arrays are way too expensive Butler matrix and Rotman lens are way too lossy

low-loss "planar" TEM Luneburg lens for 1D scan:

 $\rightarrow$  30 GHz planar TE mode air/metal Luneburg lens :









[ C. Hua et al., IEEE Trans. MTT, vol. 61, no. 1, January 2013, pp. 436-443 ]



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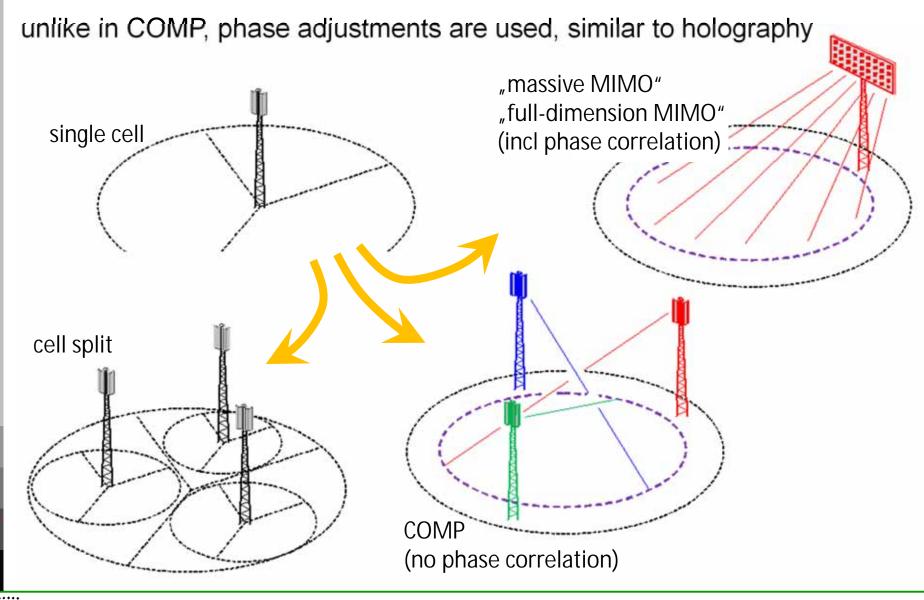
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#### - what's next



## What's next — massive MIMO





# What's next — massive MIMO

massive MIMO antenna consequences:

- large number of radiator columns cost & weight becomes more important
- phase synchronization tricky current use of compact & dense panels

a research topic at its beginnings:

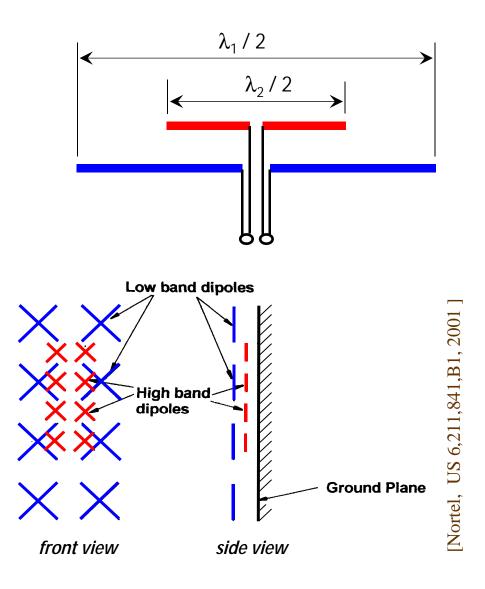
- ok for TD systems, but possible at all for FD systems?
- wide & sparse panels or fully covered cell circumference much (?) better
- can antennas support synchronization ?
- can non-synchronized repeaters reduce path correlations ?



## What's next — connected arrays

problem: cellular covers a 4:1 frequency range, but it is useless to develop 4:1 transceivers and 4:1 radiators, because array element spacing must be about  $0.6 \lambda_0$ 

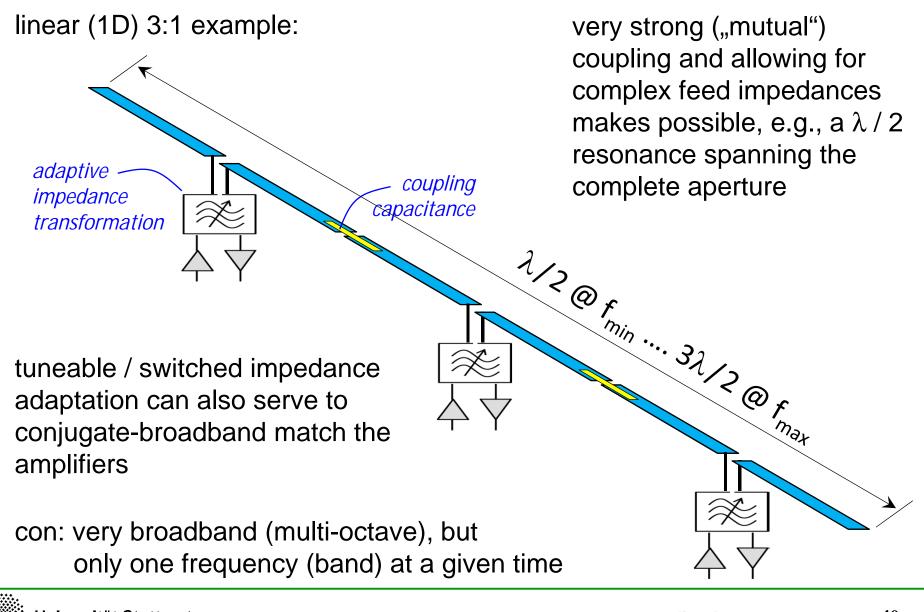
solution 1: Nortel's dual-band array



#### solution 2: connected array



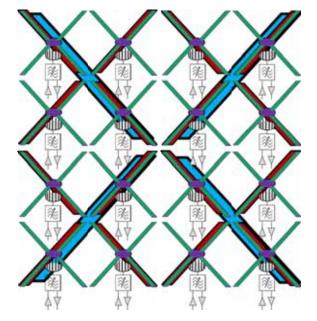
## What's next — connected arrays



# What's next — connected arrays

planar (2D) 4:1 example: identical geometrical aperture area over frequency

common feeds @  $f_0$  , 2  $f_0$  , 4  $f_0$ 



maximum directivity & maximum beamforming capability at very different (here: 4:1) frequencies from a given, common aperture



# What's next — GBps-speed mm-wave UE connections

scenario:

mm-wave directed beams using switched-beam hemispherical hotspot

multi-beam hotspot

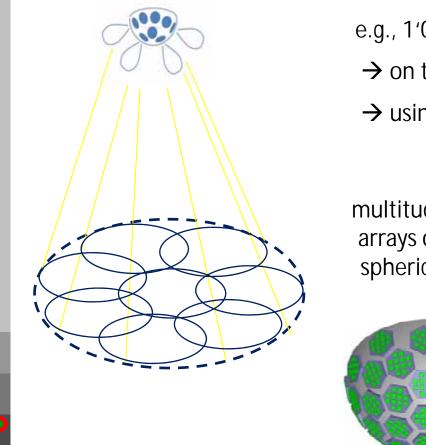




# What's next — GBps-speed mm-wave UE connections

scenario:

mm-wave directed beams using switched-beam hemispherical hotspot



e.g., 1'000 beams of 32 dBi :

- → on the surface of a sphere:  $\emptyset_{\text{sphere}} \approx 280 \, \lambda$
- → using a graded lens:  $\emptyset_{\text{sphere}} \approx \underline{14 \lambda}$

multitude of patch arrays on a hemispherical surface principle of Luneburg lens modified Luneburg lens allowing for planar feed array

